

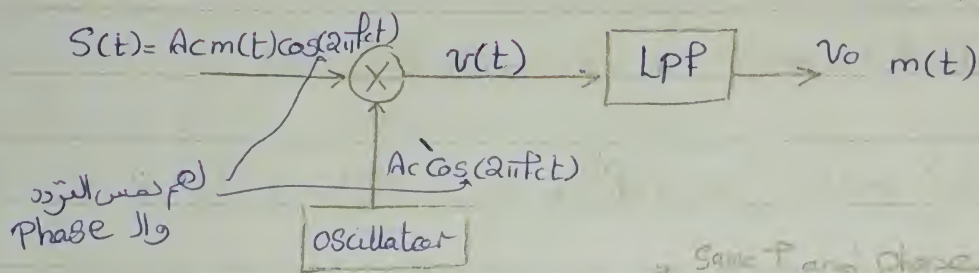


* Demodulation of D.S.B.S.C. :

أي عملية استخلاص $m(t)$ من $S(t)$.

1) Coherent demodulation :

Coherent means that the local oscillator signal is sync. with $S(t)$.
في هذه الحالة سوف نستخدم oscillator له خرج له نفس التردد و ال phase مثل $c(t)$



* assume that oscillator's o/p is coherent & sync. to the modulated carrier

let $S(t) = A_c m(t) \cos(2\pi f_c t)$

$$v(t) = S(t) * A_c \cos(2\pi f_c t)$$

$$= (A_c m(t) \cos(2\pi f_c t)) (A_c \cos(2\pi f_c t))$$

$$= A_c A_c \cdot m(t) \cdot \cos^2(2\pi f_c t)$$

$$= A_c A_c \cdot \frac{m(t)}{2} (1 + \cos(4\pi f_c t))$$

ضعف التردد

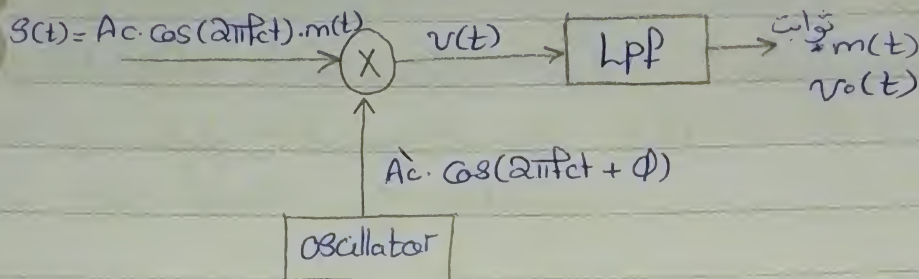
$$v(t) = \underbrace{\frac{A_c A_c m(t)}{2}}_{\text{مطلوب}} + \underbrace{\frac{A_c A_c m(t)}{2} \cos(4\pi f_c t)}_{\text{غير مطلوب}}$$

After Lpf

$$v_o = \frac{A_c A_c}{2} \cdot m(t)$$



+ If the local oscillator's output is not exactly coherent with $c(t)$ & has a phase shift ϕ from the carrier phase.



$$v(t) = A_c \cos(2\pi f_c t) \cdot m(t) * A_c \cos(2\pi f_c t + \phi)$$

$$= \frac{A_c \cdot A_c}{2} \cdot m(t) [\cos(\phi) + \cos(4\pi f_c t + \phi)]$$

$$= \underbrace{\frac{A_c \cdot A_c}{2} m(t) \cos(\phi)}_{\text{الجزء المطلوب الذي يسوي } m(t) \text{ وهديه ب Lpf}} + \underbrace{\frac{A_c \cdot A_c}{2} m(t) \cos(4\pi f_c t + \phi)}_{\text{الجزء المرفوض الذي سيتم إزالته ال Lpf}}$$

$$\cos(m) \cos(n) = \frac{1}{2} [\cos(m-n) + \cos(m+n)]$$

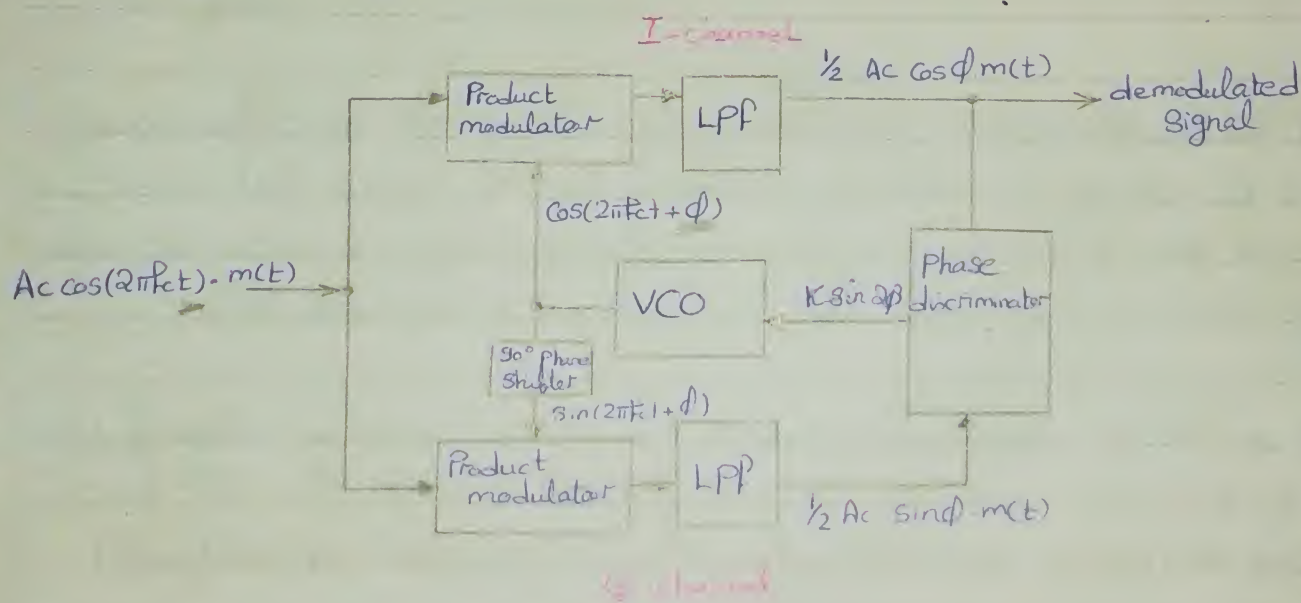
$$v_0(t) = \frac{A_c \cdot A_c}{2} m(t) \cos(\phi) \rightarrow \phi = 90^\circ \rightarrow v_0(t) = 0$$

مسألة
Square loop



2 Costas Receiver

- Practical Synchronous receiving system "Sync. means that it tends to make the oscillator's o/p phase equal to $s(t)$ phase".
- The system contains two coherent detectors
 - phase discriminator "multiplier + LPF"
- A Voltage-Controlled Oscillator (VCO) is a device whose output frequency depends on its input voltage.



- First, if the oscillator's o/p is of the same phase as the incoming DSBTC wave then, the I-channel will have the desired $m(t)$ signal and the Q-channel will have zero signal

I-channel

$$Ac \cos(2\pi fct) \cdot m(t) \cdot \cos(2\pi fct) = \frac{Ac}{2} m(t) (1 + \cos(4\pi fct))$$

$$= \underbrace{\frac{Ac}{2} m(t)}_{\text{Pass through LPF}} + \frac{Ac}{2} m(t) \cos(4\pi fct)$$



Q-channel

$$A_c \cos(2\pi f_c t) m(t) \sin(2\pi f_c t) = \frac{A_c \cdot m(t)}{2} [\sin(0) + \sin(4\pi f_c t)]$$

$$= \frac{A_c \cdot m(t)}{2} \sin(4\pi f_c t)$$

نن تفرمن ال LPP

∴ o/p of LPP will be zero.

- The oscillator cos signal is phase shifted by 90° so, it is shifted into a sine wave, a signal and its 90° shifted signal are called in phase Quadrature

Phase

- IF, the oscillator's o/p is shifted by a small angle ϕ , the Costas receiver tends to reduce this shift by using a -ve F.B. system that takes part of the I-channel o/p & the Q-channel o/p into phase discriminator which produces a dc signal that is proportional to the phase shift

- This dc value automatically controls this phase error and reduce it.

I-channel

$$\cos \phi = \cos 0 = \frac{1}{2} (\cos \phi + \cos \phi)$$

$$A_c \cos(2\pi f_c t) m(t) \cdot \cos(2\pi f_c t + \phi) = \frac{A_c \cdot m(t)}{2} [\cos(\phi) + \cos(4\pi f_c t + \phi)]$$

through LPP

Q-channel

$$\sin \phi = \cos 0 = \frac{1}{2} [\sin(\phi) + \sin(\phi)]$$

$$A_c \cos(2\pi f_c t) m(t) \cdot \sin(2\pi f_c t + \phi) = \frac{A_c \cdot m(t)}{2} [\sin(\phi) + \sin(4\pi f_c t + \phi)]$$

through LPP

AE

phase discriminator

$$\frac{A_c \cdot m(t)}{2} \cos \phi \times \frac{A_c \cdot m(t)}{2} \sin \phi = \frac{A_c^2 m^2(t)}{4} \cdot \frac{1}{2} [\sin(0) + \sin 2\phi]$$

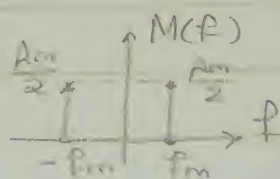


$$= \frac{A_c^2 m^2(t)}{8} \sin 2\phi$$

→ this term will pass through LPF
and dc Part will only pass.
and the o/p of it will be
 $K \cdot \sin(2\phi)$.

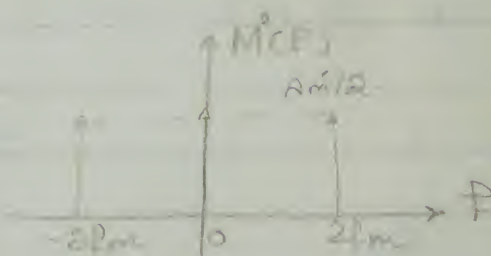
Note:

assume $m(t) = A_m \cos(2\pi f_m t)$



$$m^2(t) = \frac{A_m^2}{2} (1 + \cos 4\pi f_m t)$$

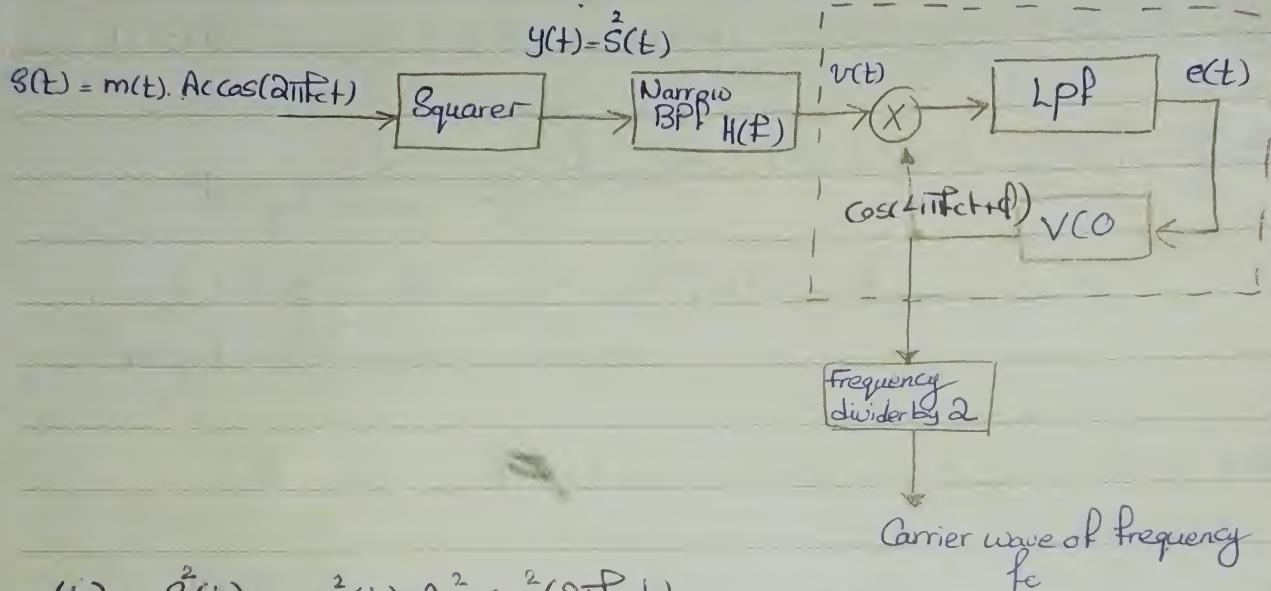
$$\therefore M^2(f) = \frac{A_m^2}{2} \delta(f) + \frac{A_m^2}{2} [\delta(f - 2f_m) + \delta(f + 2f_m)]$$



→ LPF will pass the delta component at $f=0$ which will be
 $K \sin(2\phi)$.

Demod. of DSBSCSquaring loop

الهدف من هذا عمل Carrier لا نفس ال ϕ و ال Phase من ال Carrier
في الإشارة المستقبلة لـ Coherent detection لا \rightarrow لا \rightarrow لا



$$y(t) = S^2(t) = m^2(t) A_c^2 \cos^2(2\pi f_c t)$$

$$= \frac{A_c^2}{2} m^2(t) [1 + \cos(4\pi f_c t)]$$

will pass

- after the BPF which is centered at $2f_c$.

$$v(t) = \frac{A_c^2}{2} E A F \cos(4\pi f_c t)$$

$$v(t) \cdot \cos(4\pi f_c t + \phi)$$

$$= \frac{A_c^2}{2} E A F \cdot \frac{1}{2} [\cos(\phi) + \cos(8\pi f_c t + \phi)]$$

↓ through Lpf

$$e(t) = K \cdot \cos \phi$$

which will adjust the VCO until its o/p is $\cos(4\pi f_c t)$ → has $F = 2f_c$

- The desired freq. is then f_c after division by two.